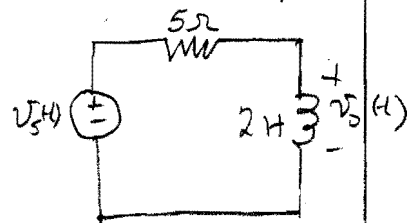
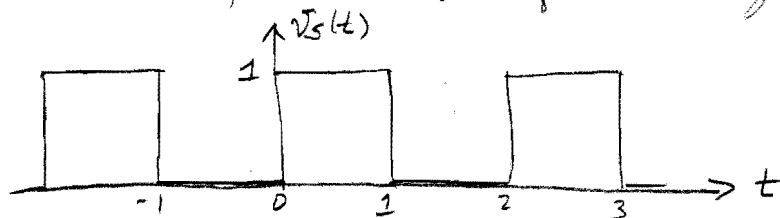


Example 17.6

Determine $v_o(t)$ in the circuit below if the input source voltage is the square wave of Example 17.1.



To analyze this circuit for $v_o(t)$, we note that while $v_s(t)$ is a steady state periodic function, we cannot use sinusoidal steady state (phasor) analysis since v_s is not a sinusoid.

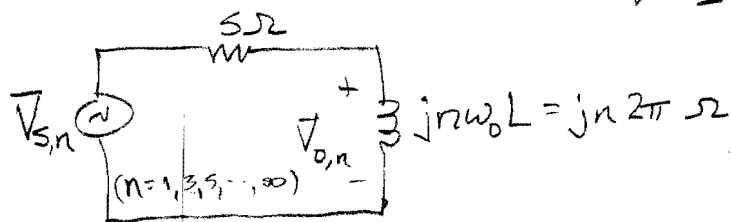
A very useful approach here is to expand $v_s(t)$ in a Fourier Series, which we did in Example 17.1.

$$v_s(t) = \frac{1}{2} + \frac{2}{\pi} \sum_{\substack{n=1 \\ (n \text{ odd})}}^{\infty} \frac{1}{n} \sin(n\pi t) \quad \text{V} \quad (1), (17.1.7)$$

Because the circuit is linear, we can use superposition. We note that (1) has a DC component and a sum of sinusoids of different frequencies $n\omega_0 = n\pi$. Each of these "harmonics" can be used as a phasor input voltage to the phasor domain circuit and we can solve for the phasor output voltage at that frequency.

- DC analysis Replace L w/ short circuit $\Rightarrow V_o|_{DC} = 0V$.
 $V_{s|DC} = \frac{1}{2} V$

- AC analysis The phasor domain circuit at frequency $\omega = n\omega_0 = n\pi$ rad/s:



From (1)

$$v_{s,n}(t) = \frac{2}{n\pi} \sin(n\pi t) \text{ V} = \frac{2}{n\pi} \cos(n\pi t - 90^\circ) \text{ V}$$

$$\bar{V}_{s,n} = \frac{2}{n\pi} \angle -90^\circ \text{ V} = \frac{-j2}{n\pi} \text{ V}$$

By voltage division, $\bar{V}_{o,n} = \frac{jn2\pi}{5 + jn2\pi} \cdot \frac{-j2}{n\pi} = \frac{4}{5 + jn2\pi} \text{ V}$

So this output voltage will change depending on the frequency of the harmonic that is input from the source.

In magnitude-angle form:

$$\bar{V}_{o,n} = \frac{4}{|5 + jn2\pi|} \angle -\tan^{-1}\left(\frac{n2\pi}{5}\right)$$

$$= \sqrt{(5 + jn2\pi) \cdot (5 - jn2\pi)}$$

$$\rightarrow |5 + jn2\pi| = \sqrt{(5 + jn2\pi)(5 - jn2\pi)} = \sqrt{25 + n^2 4\pi^2}$$

$$\therefore \bar{V}_{o,n} = \frac{4}{\sqrt{25 + 4n^2\pi^2}} \angle -\tan^{-1}\left(\frac{n2\pi}{5}\right)$$

This is the phasor output voltage for the sinusoidal steady state input voltage at the frequency of $n\omega_0 = n\pi$ rad/s.

In the time domain, notice!

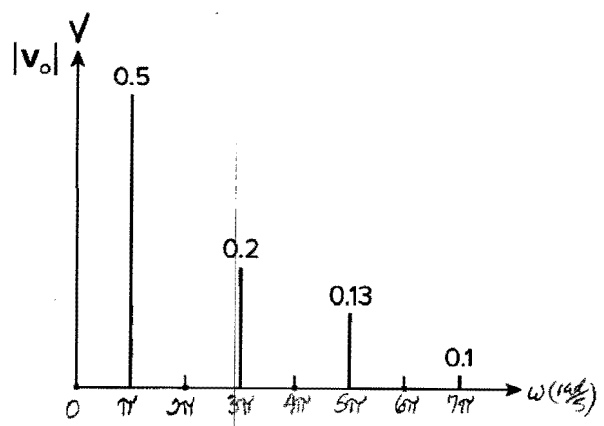
$$v_{o,n}(t) = \text{Re} \left\{ \bar{V}_{o,n} e^{j\omega_n t} \right\} = \text{Re} \left\{ \frac{4}{\sqrt{25 + 4n^2\pi^2}} \angle -\tan^{-1}\left(\frac{2n\pi}{5}\right) \cdot e^{j\omega_n t} \right\}$$

$$\therefore v_{o,n}(t) = \frac{4}{\sqrt{25 + 4n^2\pi^2}} \cos \left[n\pi t - \tan^{-1}\left(\frac{2n\pi}{5}\right) \right] \quad (n \text{ odd})$$

Because this circuit is linear we can use superposition and construct the total voltage - the actual solution - by summing all of these harmonics together w/ the dc soln:

$$v_o(t) = \sum_{\substack{n=1 \\ (n \text{ odd})}}^{\infty} \frac{4}{\sqrt{25 + 4n^2\pi^2}} \cos \left[n\pi t - \tan^{-1}\left(\frac{2n\pi}{5}\right) \right] \quad V$$

Amplitude Spectrum - Fig. 17.21



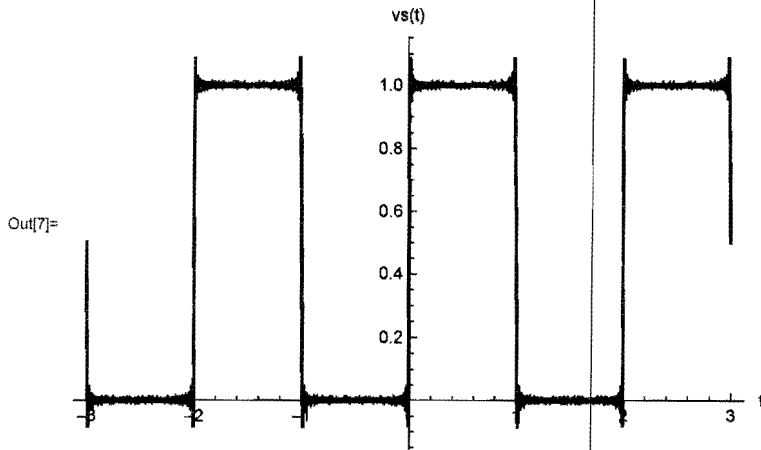
Add Example 17.11

Whites EE 221 - Circuits II

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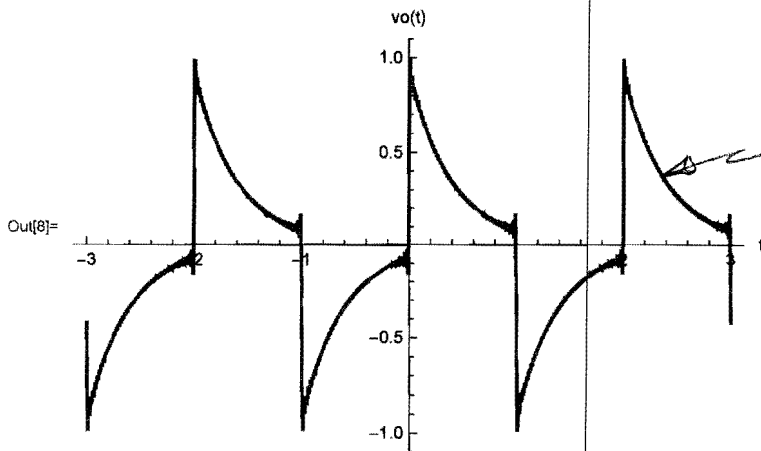
In[5]= vs[nmax_, t_] := 1/2 + 2/Pi * Sum[1/n * Sin[n*Pi*t], {n, 1, nmax, 2}]
      vo[nmax_, t_] :=
      Sum[4/Sqrt[25 + 4*n^2 * Pi^2] * Cos[n*Pi*t - ArcTan[2*n*Pi/5]], {n, 1, nmax, 2}]
In[7]= Plot[vs[101, t], {t, -3, 3}, AxesLabel -> {"t", "vs(t)"}]
  
```

n = odd



```

In[8]= Plot[vo[101, t], {t, -3, 3}, AxesLabel -> {"t", "vo(t)"}]
  
```



*makes sense: This is an STC
 CR = L/R = 2/5 = 0.4 sec.*