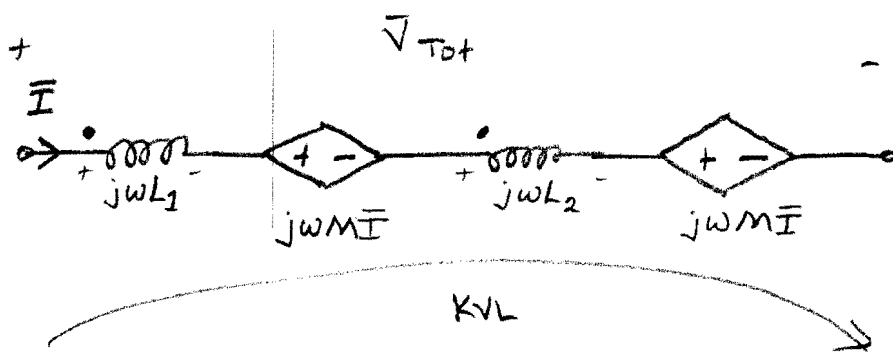


Prob. 5.1 Derive (13.18) in the text.



Substitute equivalent circuit model for the mutual coupling.



By KVL: 
$$\begin{aligned}\bar{V}_{Tot} &= j\omega L_1 \bar{I} + j\omega M \bar{I} + j\omega L_2 \bar{I} + j\omega M \bar{I} \\ &= j\omega (L_1 + M + L_2 + M) \bar{I}\end{aligned}$$

an effective inductance

$$\underline{\underline{L_{eff} = L_1 + L_2 + 2M}}$$

**Solution 13.3**

$$L_1 + L_2 + 2M = 500 \text{ mH} \quad (1), (13.18)$$

$$L_1 + L_2 - 2M = 300 \text{ mH} \quad (2), (13.19)$$

Adding (1) and (2),

$$2L_1 + 2L_2 = 800 \text{ mH} \quad (3)$$

But,  $L_1 = 3L_2$ ; or  $8L_2 = 800$ , and  $L_2 = 100 \text{ mH}$

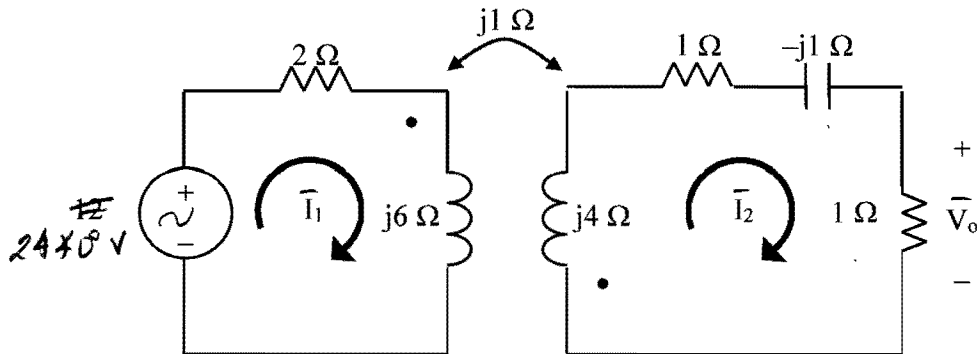
$$\underline{L_1 = 3L_2 = 300 \text{ mH}}$$

From (2),  $\overset{300}{150} + \overset{100}{50} - 2M = \overset{300}{150}$  leads to  $\underline{M = 50 \text{ mH}}$

(13.36):  $\underline{k = M/\sqrt{L_1 L_2} = 50/\sqrt{100 \times 300} = \underline{0.2887}}$

**Solution 13.7** . Determine  $\bar{V}_o$

We apply mesh analysis to the circuit as shown below.



For mesh 1,

$$(2+j6)\bar{I}_1 + j\bar{I}_2 = 24$$

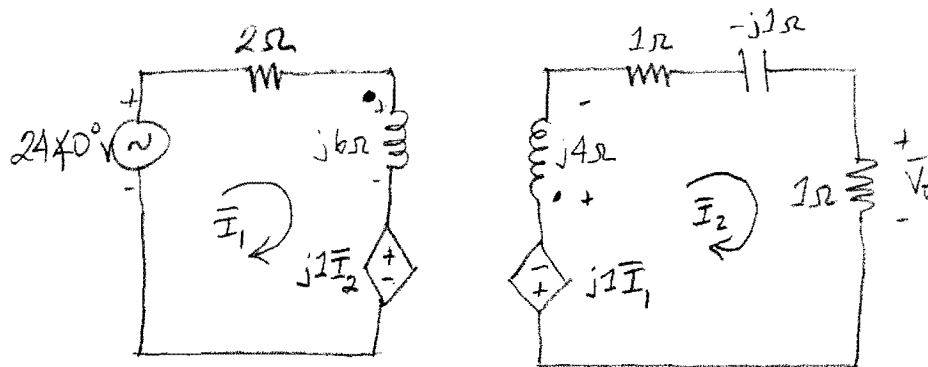
For mesh 2,

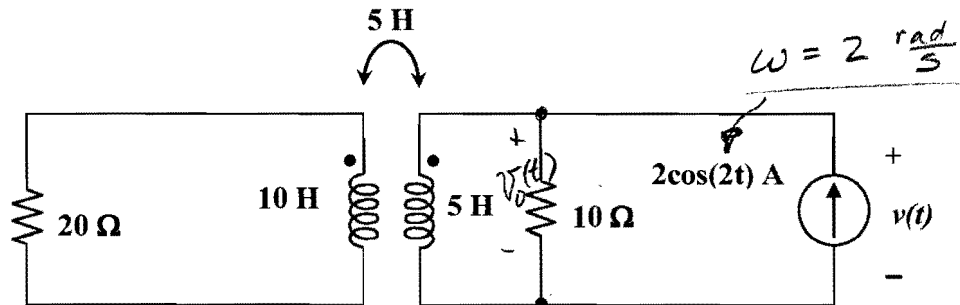
$$j\bar{I}_1 + (2-j+j4)\bar{I}_2 = j\bar{I}_1 + (2+j3)\bar{I}_2 = 0 \text{ or } \bar{I}_1 = (-3+j2)\bar{I}_2$$

Substituting into the first equation results in  $\bar{I}_2 = (-0.8762+j0.6328)$  A.

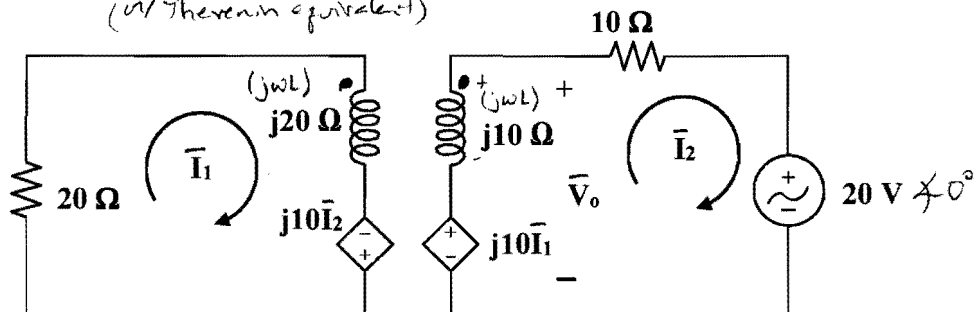
$$\underline{\bar{V}_o} = \bar{I}_2 \times 1 = \underline{1.081 \angle 144.16^\circ \text{ V.}}$$

Equivalent circuit:



**Solution 13.8**Find  $v(t)$  for the circuit in Fig. 13.77.Figure 13.77  
For Prob. 13.8.**Solution**

Step 1. We ~~need to~~ transform the circuit into the frequency domain <sup>then</sup> and replace the coupled inductors with the dependent source model. In addition, we ~~need to~~ replace the current source in parallel with the resistor with the equivalent voltage source in series with the resistor (source transformation).  
(or Thevenin equivalent)



The 10 H inductor becomes  $j2 \times 10 = j20 \Omega$  and the 5 H mutual coupling and 5 H inductor becomes  $j2 \times 5 = j10 \Omega$ . The source transformation converts the current source resistor combination of  $10 \Omega$  in parallel with the 2 A source with a  $10 \Omega$  resistor in series with a 20 V source.

Loop 1,  $20\bar{I}_1 + j20\bar{I}_1 - j10\bar{I}_2 = 0$  and loop 2,  $-j10\bar{I}_1 + j10\bar{I}_2 + 10\bar{I}_2 + 20 = 0$ . Finally  $V_o = -j10\bar{I}_2 + j10\bar{I}_1$ . Note we are representing the frequency domain value of  $v(t)$  by  $V_o$ .

Step 2. From the first loop equation we <sup>find</sup> ~~get~~  $(20 + j20)\bar{I}_1 = j10\bar{I}_2$  or  $\bar{I}_1 = (0.25 + j0.25)\bar{I}_2$ .

Sub ②: This leads to  $-j10(0.25 + j0.25)\bar{I}_2 + (10 + j10)\bar{I}_2 = -20 = (12.5 + j7.5)\bar{I}_2$  or  
 $\bar{I}_2 = 20 \angle 180^\circ / (14.5774 \angle 30.964^\circ) = 1.37199 \angle 149.036^\circ$  and

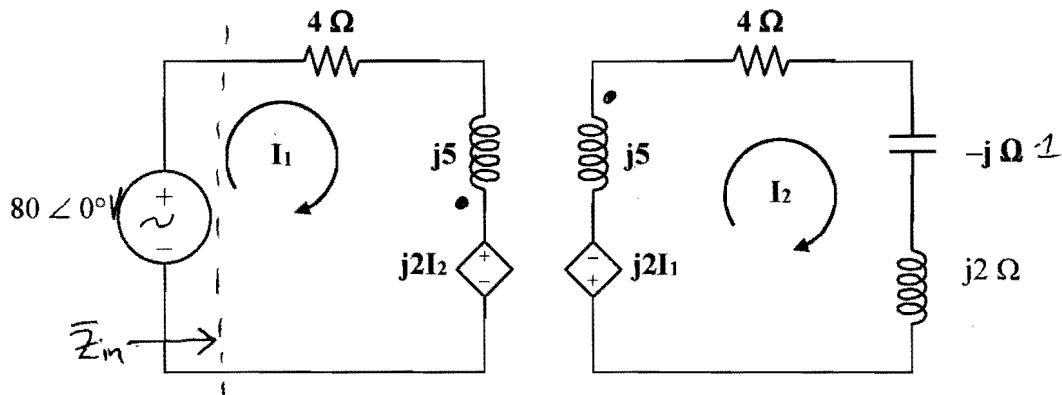
Sub ③:  $\bar{I}_1 = (0.35355 \angle 45^\circ)(1.37199 \angle 149.039^\circ) = 0.48507 \angle -165.961^\circ$ . Finally,

Sub ④:  $V_o = -j10(-1.17647 + j0.70589) + j10(-0.47058 - j0.117669)$   
 $= 8.2356 + j7.0589 = 10.847 \angle 40.6^\circ \text{ V or}$

$v(t) = 10.847 \cos(10t + 40.6^\circ) \text{ A.}$

Solution 13.13 Determine impedance seen by source.

Equivalent Circuit:



$$\text{Mesh 1: } -80 + (4+j5)I_1 + j2I_2 = 0 \text{ or } (4+j5)I_1 + j2I_2 = 80 \quad (1)$$

$$\text{Mesh 2: } j2I_1 + (4+j6)I_2 = 0 \text{ or } I_2 = [-j2/(7.2111\angle56.31^\circ)]I_1 = (0.27735\angle-146.31^\circ)I_1 \quad (2)$$

$$\text{Sub into (1): } [4+j5 + j2(-0.230769-j0.153846)]\bar{I}_1 = [4+j5+0.307692-j0.461538]\bar{I}_1 = 80$$

$$[4.307692+j4.538462]\bar{I}_1 = 80 \text{ or } \bar{I}_1 = 80/(6.2573\angle46.494^\circ) \\ = 12.78507\angle-46.494^\circ \text{ A.}$$

$$\bar{Z}_{in} = \frac{V_s}{\bar{I}_1}$$

$$\bar{Z}_{in} = 80/\bar{I}_1 = 6.2573\angle46.494^\circ \Omega = \underline{\underline{(4.308 + j4.538) \Omega}}$$

An alternate approach would be to use the equation,

$$Z_{in} = 4 + j(5) + \frac{4}{j5 + 4 - j + j2} = 4 + j5 + \frac{4}{7.2111\angle56.31^\circ} \\ = 4 + j5 + 0.5547\angle-56.31^\circ = 4 + 0.30769 + j(5 - 0.46154)$$

$$= [4.308 + j4.538] \Omega.$$

## Solution 13.24

(a)  $k = M/\sqrt{L_1 L_2} = 1/\sqrt{4 \times 2} = 0.3535$

(b)  $\omega = 4$  convert circuit to the frequency domain

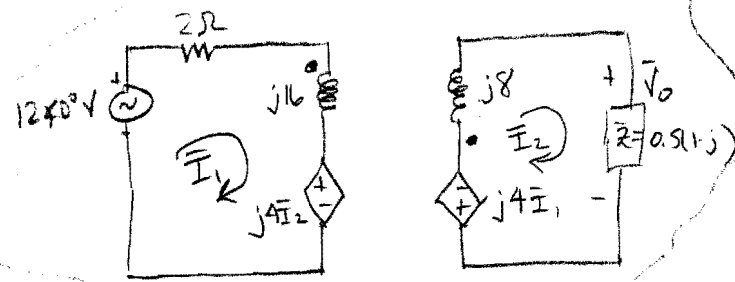
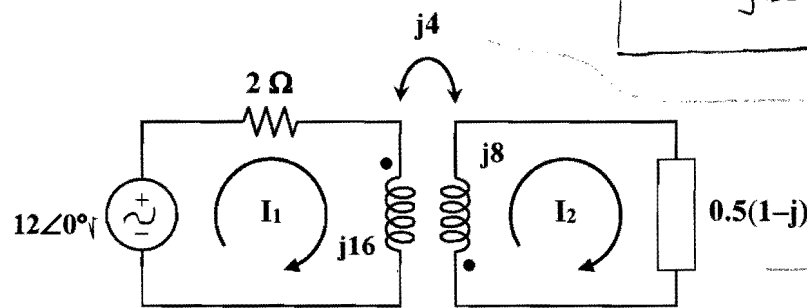
C:  $1/4 \text{ F}$  leads to  $1/(j\omega C) = -j/(4 \times 0.25) = -j$

R||C:  $1||(-j) = -j/(1-j) = 0.5(1-j)$

M:  $1 \text{ H}$  produces  $j\omega M = j4$

L:  $4 \text{ H}$  produces  $j16$

L:  $2 \text{ H}$  becomes  $j8$



KVL, mesh 1:  $12 = (2 + j16)\bar{I}_1 + j4\bar{I}_2$

or  $6 = (1 + j8)\bar{I}_1 + j2\bar{I}_2$  (1)

KVL, mesh 2:  $0 = (j8 + 0.5 - j0.5)\bar{I}_2 + j4\bar{I}_1$  or  $\bar{I}_1 = (0.5 + j7.5)\bar{I}_2/(-j4)$  (2)

$= (-1.875 + j0.125)\bar{I}_2$

Substituting (2) into (1),

$6 = (1 + j8)(-1.875 + j0.125)\bar{I}_2 + j2\bar{I}_2$

or,

$24 = (-11.5 - j51.5)\bar{I}_2$  or  $\bar{I}_2 = -24/(11.5 + j51.5) = -0.455 \angle -77.41^\circ = 0.455 \angle 102.59^\circ \text{ A}$

$\bar{V}_0 = \bar{I}_2(0.5)(1-j) = 0.3217 \angle 57.59^\circ \text{ V}$

$v_0^{(t)} = \underline{321.7 \cos(4t + 57.6^\circ) \text{ mV}}$

$v_0(t) = \text{Re}\{\bar{V}_0 e^{j\omega t}\}$

(c) From (2),  $\bar{I}_1 = (0.5 + j7.5)\bar{I}_2 / (-j4) = 0.855 \angle -81.21^\circ \text{ A}$   $\rightarrow i_1(t) = \text{Re}\{\bar{I}_1 e^{j\omega t}\}$

$i_1(t) = 0.885 \cos(4t - 81.21^\circ) \text{ A}$ ,  $i_2(t) = 0.455 \cos(4t + 77.41^\circ) \text{ A}$

At  $t = 2\text{ s}$ ,

$4t = 8 \text{ rad} = 98.37^\circ$   $\xrightarrow{180^\circ / \pi \text{ rad}}$  then  $-360^\circ$

$i_1|_{t=2\text{s}} = 0.885 \cos(98.37^\circ - 81.21^\circ) = 0.8169 \text{ A}$

$i_2|_{t=2\text{s}} = 0.455 \cos(98.37^\circ + 77.41^\circ) = -0.4249$

Eqn (13.32 w/ + sign  
both currents assumed  
entering terminals:

$w = 0.5L_1 i_1^2 + 0.5L_2 i_2^2 + M i_1 i_2$

$= 0.5(4)(0.8169)^2 + 0.5(2)(-0.4249)^2 + (1)(0.8169)(-0.4249) = 1.168 \text{ J}$

**Solution 13.36**

Following the two rules in section 13.5, we obtain the following:

$$\begin{aligned} N_2 &= n \\ N_1 &= 1 \end{aligned}$$

$$\frac{N_2}{N_1} = n$$

- (a)  $\bar{V}_2/\bar{V}_1 = -n, \quad \bar{I}_2/\bar{I}_1 = -1/n$
- (b)  $\bar{V}_2/\bar{V}_1 = -n, \quad \bar{I}_2/\bar{I}_1 = -1/n$
- (c)  $\bar{V}_2/\bar{V}_1 = n, \quad \bar{I}_2/\bar{I}_1 = 1/n$
- (d)  $\bar{V}_2/\bar{V}_1 = n, \quad \bar{I}_2/\bar{I}_1 = -1/n$



**Solution 13.37**

A 240/2400 V (rms) step-up ideal transformer delivers 50 kW to a resistive load. Calculate: (a) the turns ratio, (b) the primary current, (c) the secondary current.

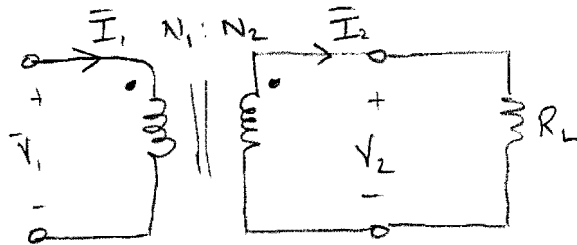
**Solution**

(a)  $n = \frac{V_2}{V_1} = \frac{2400}{240} = 10$

(b)  $S_1 = V_1(I_1)^* = S_2 = V_2(I_2)^* = 50,000^{\text{kVA}}$  which leads to

$$I_1 = 50,000/240 = 208.3 \text{ A.}$$

(c)  $I_2 = 50,000/2,400 = 20.83 \text{ A.}$



**Solution 13.42**

For the circuit in Fig. 13.106, determine the power absorbed by the  $2\ \Omega$  resistor. Assume the  $120\ \text{V}$  source is an rms value.

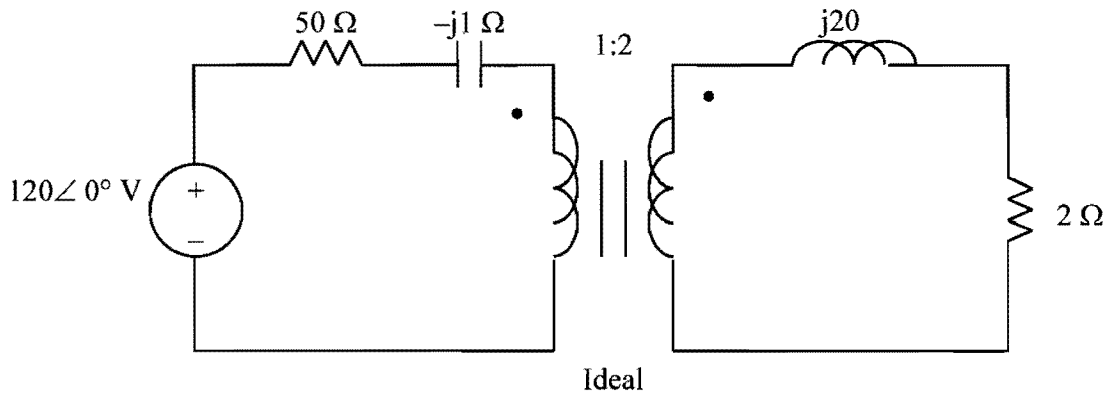
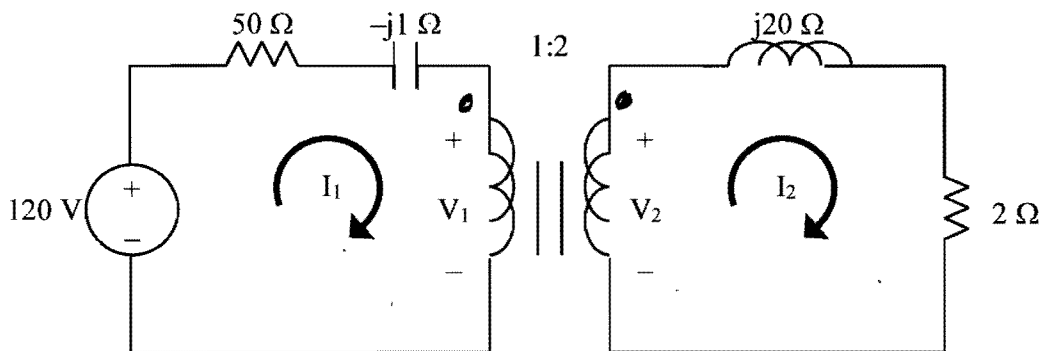


Figure 13.106  
For Prob. 13.42.

**Solution**

We apply mesh analysis to the circuit as shown below.



For mesh 1,

$$-120 + (50 - j)\bar{I}_1 + \bar{V}_1 = 0 \quad (1)$$

For mesh 2,

$$-\bar{V}_2 + (2 + j20)\bar{I}_2 = 0 \quad (2)$$

At the transformer terminals,

$$\bar{V}_2 = 2\bar{V}_1 \text{ or } 2\bar{V}_1 - \bar{V}_2 = 0 \quad (3)$$

$$\bar{I}_1 = 2\bar{I}_2 \text{ or } \bar{I}_1 - 2\bar{I}_2 = 0 \quad (4)$$

From (1) to (4),

$$\begin{bmatrix} 50-j & 0 & 1 & 0 \\ 0 & 2+j20 & 0 & -1 \\ 0 & 0 & 2 & -1 \\ 1 & -2 & 0 & 0 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} 120 \\ 0 \\ 0 \\ 0 \end{bmatrix}$$

Solving this with MATLAB,

```
>> A = [(50-j) 0 1 0; 0 (2+20j) 0 -1; 0 0 2 -1; 1 -2 0 0]
```

A =

Columns 1 through 3

```
50.0000 - 1.0000i    0    1.0000
    0    2.0000 + 20.0000i    0
    0    0    2.0000
1.0000    -2.0000    0
```

Column 4

```
0
-1.0000
-1.0000
0
```

```
>> B = [120;0;0;0]
```

B =

```
120
0
0
0
```

```
>> C = inv(A)*B
```

C =

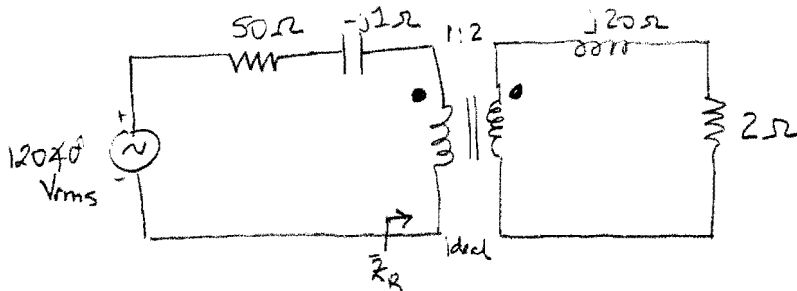
```
2.3614 - 0.8170i    (I1)
1.1806 - 0.0934i    (I2)
2.1159 + 11.7136i    (V1)
4.2318 + 23.4272i    (V2)
```

$$\bar{I}_2 = (1.1806 - j0.0934) \text{ A or } 1.1843 \angle -4.52^\circ \text{ A}$$

The power absorbed by the 2-Ω resistor is

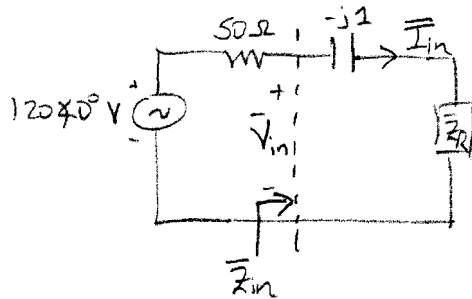
$$P = \overset{\text{RMS}}{|\bar{I}_2|^2 R} = (1.1843)^2 \cdot 2 = \underline{\underline{2.805 \text{ W}}}$$

OK... that was a long, but rigorous, sol'n. Here's the quick way to solve these types of problems.



$$\text{From (13.60)}, \bar{Z}_R = \frac{\bar{Z}_L}{n^2} = \frac{2 + j20}{(2/1)^2} = 0.5 + j5 \Omega$$

Equivalent circuit for the primary side:



$$\bar{Z}_{in} = \bar{Z}_R - j1 = 0.5 + j5 - j1 = 0.5 + j4 \Omega$$

By voltage division:

$$\begin{aligned} \bar{V}_{in} &= \frac{\bar{Z}_{in}}{\bar{Z}_{in} + 50} \cdot V_s = \frac{0.5 + j4}{0.5 + j4 + 50} \cdot 120^\circ \\ &= 9.549 \angle 78.35^\circ \text{ V} \end{aligned}$$

The time-average power delivered to the  $\bar{V}_{in}$  part is (w/  $\bar{V}_{in}$  in rms)

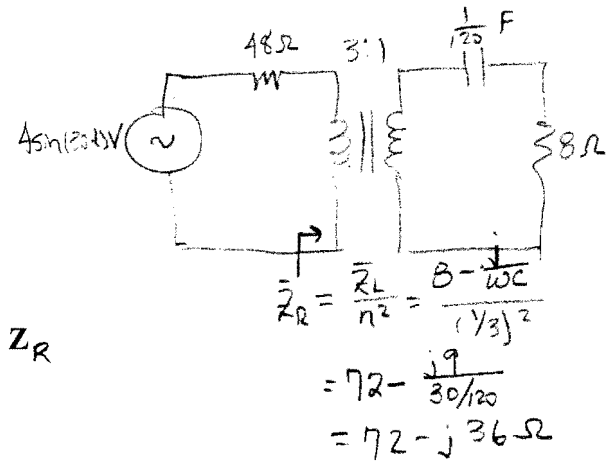
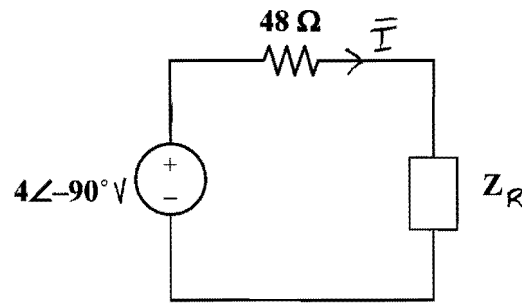
$$P_{in} = \text{Re}(\bar{V}_{in} \bar{I}_{in}^*) = \text{Re}\left\{ \frac{|\bar{V}_{in}|^2}{\bar{Z}_{in}^*} \right\} = |\bar{V}_{in}|^2 \text{Re}\left\{ \frac{1}{\bar{Z}_{in}^*} \right\} = 9.549^2 \text{Re}\left\{ \frac{1}{0.5 - j4} \right\}$$

$$= 91.184 \text{Re}\{0.03077 + j0.2462\} = 91.184 \cdot 0.03077 = 2.806 \text{ W}$$

Because the components from the input port  $\bar{V}_{in}$  to the  $2 \Omega$  load is comprised of lossless components, all of this time average power is ultimately delivered to the  $2 \Omega$  load.

$$\therefore \underline{\underline{P_{2\Omega} = P_{in} = 2.806 \text{ W}}}$$

## Solution 13.45



$$\bar{Z}_L = 8 - \frac{j}{\omega C} = 8 - j4, \quad n = 1/3$$

$$\bar{Z}_R = \frac{Z_L}{n^2} = 9Z_L = 72 - j36 \quad \Delta$$

$$\bar{I} = \frac{4\angle -90^\circ}{48 + 72 - j36} = \frac{4\angle -90^\circ}{125.28\angle -16.7^\circ} = 0.03193\angle -73.3^\circ \text{ A}$$

$48 + \bar{Z}_R$

We now have some choices, we can go ahead and calculate the current in the second loop and calculate the power delivered to the 8-ohm resistor directly, or we can merely say that the power delivered to the equivalent resistor in the primary side must be the same as the power delivered to the 8-ohm resistor. Therefore,

*time-average*  
*time-average*

$$\underline{P}_{8\Omega} = \frac{|\bar{I}|^2}{2} \cdot 72 = 0.5098 \times 10^{-3} \cdot 72 = \underline{\underline{36.71 \text{ mW}}}$$

The student is encouraged to calculate the current in the secondary and calculate the power delivered to the 8-ohm resistor to verify that the above is correct.

**Solution 13.55**

For the circuit in Fig. 13.119, calculate the equivalent resistance.

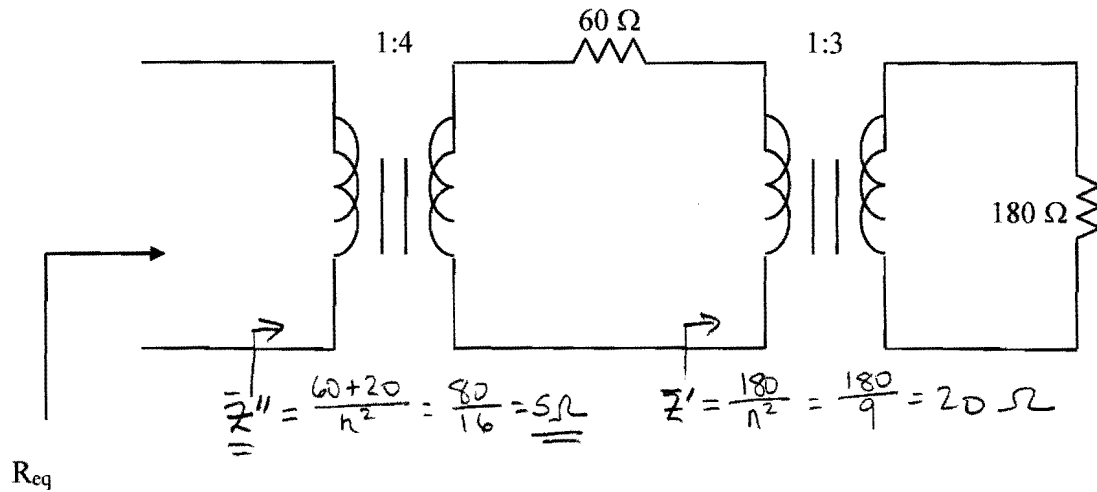


Figure 13.119  
For Prob. 13.55.

**Solution**

We first reflect the 80- $\Omega$  resistance to the middle circuit.

$$Z' = 60 + [180/(3)^2] = 60 + 20 = 80\ \Omega$$

We now reflect this to the primary side.

$$\underline{\underline{R_{eq} = Z'/(4)^2 = 5\ \Omega.}}$$